# applied optics

### Calibration of oblique-incidence reflectivity difference for label-free detection of a molecular layer

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Oblique-incidence reflectivity difference (OI-RD) is a form of polarization-modulation ellipsometry that measures properties of thin films on a solid surface through the change in polarization state of light upon reflection from the surface. The measurement accuracy depends on the precision of the phase modulation amplitude and azimuthal alignments of key polarizing optical elements and, thus, requires careful calibration. In the present work, we describe robust methods of such calibrations that enable precise determination of the modulation amplitude and static retardation of a phase modulator and azimuths of key polarizing optics in an OI-RD system. © 2016 Optical Society of America

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#### **1. INTRODUCTION**

Ellipsometry is a class of optical techniques widely used to measure the thickness and refractive index of thin films on solid surfaces through the measurement of polarization state change. In nulling ellipsometry [1], the intensity of a polarized light beam reflected from a solid surface is extinguished through azimuthal adjustments of polarization optics in the instrument. The azimuths of the polarization optics required to null the intensity at the detector are used to extract the change in the polarization state. To detect ellipsometry signals at frequencies away from 1/f noise, one typically uses polarization-modulation ellipsometry (PME) [2], in which a photo-elastic modulator (PEM) or electro-optic phase modulator (EOM) is utilized to introduce an oscillatory phase retardation along one major axis at the frequency  $\Omega$ . As a result, the intensity of the detected ellipsometry signal consists of harmonics of  $\Omega$ , and the first and second harmonic components are routinely used to extract the polarization change with an improved signal-to-noise ratio (SNR). OI-RD is a form of nulling PME [3,4]. Reflectivities for obliquely incident *p*- and *s*-polarized light change disproportionately in response to a change in thickness or dielectric response of the film. OI-RD directly measures such disproportional phase and amplitude changes using the first and second harmonics of the ellipsometry signal. OI-RD has been successfully applied by Zhu and coworkers to studies of a wide variety of materials and processes, including ion sputtering and thermal annealing [5,6], gas adsorption [3,7], epitaxial growth [8,9], electrochemical deposition [10,11], and optical biosensing [12–17].

OI-RD usually consists of a polarizer, a PEM, a phase shifter, and an analyzer. The polarizer generates a linearpolarized incident light beam, the PEM alters the light beam from *p*-polarized to *s*-polarized at the frequency  $\Omega$ , and the phase shifter alters the static phase difference between the *p*- and *s*-polarized components. The analyzer mixes the two polarized components such that afterward the first and second harmonics of the detected signal yield information on the polarization state change. The PEM consists of a fused silica bar vibrating at a natural resonant frequency sustained by a piezoelectric transducer. The periodic stress produces an extra time-varying phase along the stressed axis. The precisions of the phase modulation amplitude and the azimuths of all of the polarizing optics affect the precision of the polarization state measurement. As a result, procedures need to be established to conveniently and precisely determine static and alternating phase retardation of the PEM and azimuths of key polarizing optics, and how precisions of these parameters quantitatively affect the extracted polarization changes.

In this work, we describe a set of calibration methods for OI-RD that (1) measure the modulation amplitude and static phase retardation of a PEM, and (2) set azimuths of the polarizer, PEM, phase shifter, and analyzer. Since a PEM is a stable device, part (1) needs to be performed only occasionally. Part (2) may be performed more frequently with the sample in place.

#### 2. DESCRIPTION OF AN OI-RD SYSTEM

Figure 1 shows the schematic diagram of an OI-RD system [13,16,18–20]. With the polarizer azimuth P at 45°, the PEM azimuth M at 0°, the phase shifter azimuth PS at 0°, the analyzer azimuth A at 135°, and the PEM phase modulation  $\varphi = \varphi_A \cos \Omega t$  ( $\Omega = 2\pi f, f = 50$  kHz), the detected ellipsometry signal I(t) is expressed as a function of the modulation frequency  $\Omega$ ,

$$I(t) = I(dc) + I(\Omega) \cos \Omega t + I(2\Omega) \cos 2\Omega t$$
 + higher terms.

The OI-RD system measures the first harmonic  $I(\Omega)$  and the second harmonic  $I(2\Omega)$  with lock-in amplifiers. If there are no errors in any of the azimuths,  $I(\Omega)$  and  $I(2\Omega)$  can be expressed as

$$I(\Omega) = -I_0 \eta_1 J_1(\varphi_A) |r_p| |r_s| \sin(\delta - \Phi_{\rm ps} - \Phi_{\rm sys}), \qquad (2)$$

(1)

$$I(2\Omega) = I_0 \eta_2 J_2(\varphi_A) |r_p| |r_s| \cos(\delta - \Phi_{\rm ps} - \Phi_{\rm sys}).$$
 (3)

 $I_0$  is the intensity of the incident light beam.  $\eta_1$  and  $\eta_2$  are frequency-dependent amplification factors for the first harmonic and the second harmonic, respectively, of the entire detection electronics, including the transducer and cables.  $\varphi_A$  is the modulation amplitude of the PEM (in radians).  $J_1(\varphi_A)$  and  $J_2(\varphi_A)$  are Bessel functions of the first kind.  $|r_p|$  and  $|r_s|$  are



**Fig. 1.** Schematic diagram of OI-RD system particular for microarray detection. Polarization optics (polarizer, photo-elastic modulator (PEM), phase shifter, and analyzer) are used to analyze the polarization state change of light reflection from a microarray. Light scanning along the *y* direction achieved by a *f*-theta lens with a scan mirror and a translation stage, scanning along the *x* direction, provide the spatial resolution for the OI-RD system. Photodiode, electronic transducer, and lock-in amplifiers are used to measure light intensity. For accurate data reduction, the azimuths for the polarizer, PEM, phase shifter, and analyzer are set at  $P = 45^\circ$ ,  $M = 0^\circ$ ,  $PS = 0^\circ$ , and  $A = 135^\circ$ , respectively.

amplitudes of reflectivity for *p*- and *s*-polarized components, and  $\delta = \Phi_p - \Phi_s$  is the phase difference between the *p*- and *s*-polarized components acquired upon reflection from the surface.  $\Phi_{ps}$  is the adjustable phase difference between the *p*- and *s*-polarization introduced by the phase shifter, and  $\Phi_{sys}$  is the phase difference between the *p*- and *s*-polarization contributed by all of the optical elements in the beam path other than the sample and the phase shifter.

Let  $\delta_0$  be the phase difference between the *p*- and *s*-polarization from the bare substrate surface. One can adjust  $\Phi_{ps}$  until  $\delta_0 - \Phi_{ps} - \Phi_{sys} = 0$ , and the first harmonic  $I(\Omega) = 0$  on the bare substrate surface. When a thin layer of molecules (thickness *d*) is subsequently added to the bare surface, the phase difference between the *p*- and *s*-polarization becomes  $\delta$ , and the change  $\Delta\delta = \delta - \delta_0$  can be determined using the resultant first harmonic and the second harmonic amplitudes;

$$I(\Omega) = -I_0 \eta_1 J_1(\varphi_A) |r_p| |r_s| \sin(\delta - \delta_0).$$
 (4)

$$I(2\Omega) = I_0 \eta_2 J_2(\varphi_A) |r_p| |r_s| \cos(\delta - \delta_0).$$
 (5)

When  $\Delta\delta$  is small, the first harmonic is used to yield  $\Delta\delta$ .  $\Delta\delta$  is related to the physical properties of the molecular layer, such as the thickness *d* as follows [13]:

$$\Delta \delta \cong \frac{-4\pi \sqrt{\epsilon_s} \cos \theta}{(\epsilon_0 - \epsilon_s)(\cot^2 \theta - \epsilon_s/\epsilon_0)} \frac{(\epsilon_d - \epsilon_o)(\epsilon_d - \epsilon_s)}{\epsilon_d} \frac{d}{\lambda}.$$
 (6)

 $\lambda$  is the wavelength of the light beam.  $e_s$ ,  $e_0$ , and  $e_d$  are the optical dielectric constants of the ambient, solid substrate, and molecular layer, respectively.  $\theta$  is the incident angle on the substrate surface bearing samples. For detection on the interface of glass and water, we set  $\theta = 37.5^{\circ}$  for a betterSNR. Thus, in OI-RD, one measures the molecular thickness d through the detection of  $\Delta\delta$ .

Yet, Eqs. (2)–(5) are valid only when azimuths are set accurately at the specified values, i.e.,  $P = 45^{\circ}$ ,  $M = 0^{\circ}$ ,  $PS = 0^{\circ}$ , and  $A = 135^{\circ}$ . In general, the first harmonic and the second harmonic depend on these azimuths as

$$I(\Omega) = I_0 \eta_1 J_1(\varphi_A) \sin 2(M - P)$$

$$\times \left\{ |r_p||r_s| \sin 2A \begin{bmatrix} \cos 2PS \cos(\delta - \Phi_{sys}) \sin \Phi_{ps} \\ -\sin(\delta - \Phi_{sys}) \cos \Phi_{ps} \end{bmatrix} \right]$$

$$- \sin 2PS \sin \Phi_{ps}[|r_p|^2 \cos^2 A - |r_s|^2 \sin^2 A] \right\}, \quad (7)$$

$$I(2\Omega) = -I_0 \eta_2 J_2(\varphi_A) \sin 2(M - P)$$

$$\times \left\{ [|r_p|^2 \cos^2 A - |r_s|^2 \sin^2 A] \right\}$$

$$\times \begin{bmatrix} \cos 2PS \sin 2(M - PS) \\ +\sin 2PS \cos 2(M - PS) \cos \Phi_{ps} \end{bmatrix} + |r_p||r_s| \sin 2A \\ \times \begin{bmatrix} \sin 2PS \sin 2(M - PS) \cos(\delta - \Phi_{sys}) \\ -\cos 2PS \cos 2(M - PS) \cos(\delta - \Phi_{sys}) \cos \Phi_{ps} \\ -\cos 2(M - PS) \sin(\delta - \Phi_{sys}) \sin \Phi_{ps} \end{bmatrix} \right\}.$$

Equations (7) and (8) show that  $|r_p|^2 \cos^2 A - |r_s|^2 \sin^2 A$ ,  $\sin(\delta - \Phi_{sys})$ , and  $\cos(\delta - \Phi_{sys})$  are generally coupled in both harmonics. If one still uses Eqs. (4) and (5) to calculate  $\Delta\delta$ , the result is no longer accurate unless the azimuths of all polarizing optics and the phase retardation from the PEM are carefully set or calibrated.

# 3. CALIBRATION OF THE PHOTO-ELASTIC MODULATOR

The PEM is a critical element in an OI-RD system. The modulation amplitude  $\varphi_A$  appears in the amplitudes of the first and the second harmonic signals. The calibration of  $\varphi_A$  can be performed in the straight-through configuration [21], as shown in Fig. 2, in which the sample is removed, and the PEM is placed between two crossed polarizers with transmission axes at 45° and 135° with respect to the modulation axis (i.e., the mechanically stressed axis). The detector is a photodiode with its output connected to an oscilloscope or to two lock-in amplifiers. For a phase modulation in the form of  $\varphi = \varphi_A \cos \Omega t$ , the light intensity at the detector is given by

$$I(t) = \frac{I_0}{2} [1 - \cos(\varphi_A \cos \Omega t)]$$
  
=  $\frac{I_0}{2} [1 - \eta_0 J_0(\varphi_A) + 2\eta_2 J_2(\varphi_A) \cos 2\Omega t$   
+ higher terms · · ·]. (9)

It consists of only even harmonics of modulation frequency. A number of features can be used to calibrate the modulation amplitude  $\varphi_A$ : (1) the distinctive waveform on the oscilloscope with  $\varphi_A$  being an integral multiple of  $\pi$  radians [22–24]; (2) the Bessel function zeros, such as  $J_0(\varphi_A)$ , being zero when  $\varphi_A$  is equal to 2.405, and  $J_2(\varphi_A)$  being zero when  $\varphi_A$  is equal to 5.136; and (3) the Bessel function maxima with  $J_1(\varphi_A)$  being maximized when  $\varphi_A$  is equal to 1.841 [25]. These are single-point calibration techniques. We propose a calibration method that involves curve fitting the entire waveform  $\cos(\varphi_A \cos \Omega t)$  and is thus more accurate.

The function  $\cos(\varphi_A \cos \Omega t)$  shows a "flat topped" and "flat bottomed" characteristic when  $\varphi_A$  is a multiple of  $\pi$  radians. By adjusting  $\varphi_A$  to give a "flat topped" or "flat bottomed" waveform by visual inspection,  $\varphi_A$  is equal to  $\pi$  within an accuracy of 1%–2% [22,26]. We find that the entire temporal waveform is sensitive to the modulation amplitude  $\varphi_A$ , and we can determine  $\varphi_A$  much more accurately by fitting the waveform to the function  $\cos(\varphi_A \cos \Omega t)$  instead of by visual inspection. We set values of modulation amplitude  $\varphi_A^i$  on the PEM controller (PEM100, Hinds Instruments, USA) and record the corresponding waveforms with a digital oscilloscope. The



**Fig. 2.** Straight-through configuration to calibrate modulation amplitude  $\varphi_A$  and static phase retardation  $\varphi_0$  of the PEM.

waveforms are then fitted to function  $\cos(\varphi_A \cos \Omega t)$  to yield the fitted modulation amplitude  $\varphi_A^f$ . Figure 3 shows the waveforms (open circles) obtained with set values  $\varphi_A^s$  and the corresponding fitting curves (solid lines) using the fitted  $\varphi_A^f$ . The fitted values  $\varphi_A^f$  are generally larger than the set values  $\varphi_A^s$ . Since the modulation amplitude  $\varphi_A$  is a linear function of the modulator voltage from the controller [27], it is reasonable that the relationship between the modulation amplitude setting value  $\varphi_A^s$  and the fitted value  $\varphi_A^f$  is linear. Figure 4 shows the linear



**Fig. 3.** Digitized oscilloscope waveforms (open circles) versus fitting waveforms (solid lines) with modulation amplitude setting values  $\varphi_A^s$  and fitting values  $\varphi_A^f$  being (a)  $\varphi_A^s = 2.513$ ,  $\varphi_A^f = 2.627$ ; (b)  $\varphi_A^s = 3.142$ ,  $\varphi_A^f = 3.230$ ; (c)  $\varphi_A^s = 4.398$ ,  $\varphi_A^f = 4.515$ .



**Fig. 4.** Linear relationship between modulation amplitude setting values  $\varphi_A^f$  and fitting values  $\varphi_A^f$ .

relationship between  $\varphi_A^f$  and  $\varphi_A^s$  to be  $\varphi_A^f = 1.011 \times \varphi_A^s + 0.059$ , with  $\varphi_A^s$  increasing from zero to  $2\pi$ .

We confirmed the accuracy of the curve-fitted  $\varphi_A$  values using a single-point calibration method and the fact that  $J_2(\varphi_A)$  is zero when  $\varphi_A$  is equal to 5.136. We performed measurements with the same setup except that we replaced the oscilloscope with lock-in amplifiers. We set  $\varphi_A^s$  from 4.9 to 5.1 in increments of 0.01 and recorded the second harmonic  $I_0\eta_2 J_2(\varphi_A)$  with the lock-in amplifier, which changed from positive to negative. By fitting the measured  $I_0\eta_2 J_2(\varphi_A^i)$  as a linear function of  $\varphi_A^s$ , we found that  $I_0\eta_2 J_2(\varphi_A^i) =$  $-70.415 \times \varphi_A^i + 353.437$ , as shown in Fig. 5. It vanishes when  $\varphi_A^s$  is 5.019. Based on the curve-fitting calibration method, this corresponds to an actual  $\varphi_A^f = 5.133$ , 0.003 or 0.06% from the value  $\varphi_A = 5.136$  that makes  $J_2(\varphi_A)$  vanish. This demonstrates the accuracy of the curve-fitting calibration method.

Equations (2) and (3) are derived assuming the phase retardation from the PEM has the form  $\varphi = \varphi_A \cos \Omega t$ . However, there exists a static birefringence in the PEM so that a static



**Fig. 5.** Linear relationship between the second harmonic amplitude with modulation amplitude  $\varphi_A^s$  around zero values of  $J_2(\varphi_A^s)$ .

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phase retardation  $\varphi_0$  is present when the PEM is not resonantly driven by an alternating transducer [28–31]. Replacing  $\varphi = \varphi_A \cos \Omega t$  with  $\varphi = \varphi_0 + \varphi_A \cos \Omega t$  to take into account this static birefringence, the detected ellipsometry signal, Eq. (9), is now expressed as

$$I(t) = \frac{I_0}{2} [1 - \cos(\varphi_0 + \varphi_A \cos \Omega t)]$$
  
= 
$$\frac{I_0}{2} \begin{bmatrix} 1 - \cos \varphi_0 \eta_0 J_0(\varphi_A) + 2 \sin \varphi_0 \eta_1 J_1(\varphi_A) \cos \Omega t + \\ 2 \cos \varphi_0 \eta_2 J_2(\varphi_A) \cos 2\Omega t + \text{higher terms} \cdots \end{bmatrix}.$$
  
(10)

For ideal modulators ( $\varphi_0 = 0$ ), there is no first harmonic in the signal [Eq. (9)]; whereas for actual modulators ( $\varphi_0 \neq 0$ ), sin  $\varphi_0$  is finite leading to a non-zero first harmonic in Eq. (10).  $\varphi_0$  can be derived from the ratio of the first harmonic amplitude ( $2 \sin \varphi_0 \eta_1 J_1(\varphi_A)$ ) to the second harmonic amplitude ( $2 \cos \varphi_0 \eta_2 J_2(\varphi_A)$ ). To simplify the analysis, we set  $\varphi_A^s = 2.544$  to have  $\varphi_A^f = 2.631$  with which  $J_1(\varphi_A^f) = J_2(\varphi_A^f)$ . With  $\eta_1/\eta_2 = 1.36$  (described in the following section), we find tan  $\varphi_0$  for our present PEM to be 0.011, which is consistent with previously reported findings [28,32].

The small static phase retardation  $\varphi_0$  introduced by the PEM used does not affect the information on the molecular layer from OI-RD signals. In the presence of a non-zero static phase retardation  $\varphi_0$ , the first and the second harmonic of the OI-RD signal can be expressed as

$$I(\Omega) = -I_0 \eta_1 J_1(\varphi_A) |r_p| |r_s| \sin(\delta - \Phi_{\rm ps} - \Phi_{\rm sys} - \varphi_0), \quad (11)$$

$$I(2\Omega) = I_0 \eta_2 J_2(\varphi_A) |r_p| |r_s| \cos(\delta - \Phi_{\rm ps} - \Phi_{\rm sys} - \varphi_0).$$
 (12)

During the nulling step, the first harmonic becomes zero on a bare substrate surface with  $\Phi_{ps} = \delta_0 - \Phi_{sys} - \varphi_0$ . Thus, the static phase retardation  $\varphi_0$  only changes the phase difference  $\Phi_{ps}$  of the phase shifter that is required to null the first harmonic signal from a bare substrate. The subsequent first harmonic is still proportional to  $\sin(\delta - \delta_0)$ , and the second harmonic amplitude of the molecular layer is proportional to  $\cos(\delta - \delta_0)$ , which are the same as the results when  $\varphi_0 = 0$  in the PEM crystal.

It is easy to see that the above calibration methods for the modulation amplitude  $\varphi_A$  and the static phase retardation  $\varphi_0$  are not affected by errors in azimuths of the polarizer and the analyzer shown in Fig. 2, while they are required to be at exactly  $P = 45^{\circ}$  and  $A = 135^{\circ}$  relative to the modulator axis in some calibration methods [23,33]. The light intensity reaching the detector in Fig. 2 is expressed as a function of azimuths P and A as

$$I(t) = I_0 \left\{ \cos^2(A - P) + \frac{\sin 2A \sin 2P}{2} [\cos(\varphi_0 + \varphi_A \cos \Omega t) - 1] \right\}$$
$$= I(dc) + I(\Omega) \cos \Omega t + I(2\Omega) \cos 2\Omega t + \text{higherterm, (13)}$$

$$I(dc) = \eta_0 I_0 \bigg\{ \cos^2(A - P) + \frac{\sin 2A \sin 2P}{2} [\cos \varphi_0 J_0(\varphi_A) - 1] \bigg\},$$
 (14)

$$I(\Omega) = -\eta_1 I_0 \sin 2A \sin 2P J_1(\varphi_A) \sin \varphi_0, \qquad (15)$$

$$I(2\Omega) = -\eta_2 I_0 \sin 2A \sin 2P J_2(\varphi_A) \cos \varphi_0.$$
 (16)

Equations (13)–(16) show that the waveform of  $\cos(\varphi_0 + \varphi_A \cos \Omega t)$ , the zero point of the second harmonic, and the ratio of the first harmonic to the second harmonic are all independent of azimuths *P* and *A*.

#### 4. CALIBRATING AZIMUTHS OF POLARIZING OPTICS

Typically, setting azimuths of polarizing optics in ellipsometry is accomplished by finding nulls of suitable signals at special azimuth values [31,32,34–36]. Similarly here, without a phase shifter ( $PS = 0^\circ$ ,  $\Phi_{ps} = 0^\circ$ ), we can express the first and the second harmonic of an OI-RD signal as a function of the azimuths as follows from Eqs. (7) and (8):

$$I(\Omega) = -I_0 \eta_1 J_1(\varphi_A) \sin 2(M - P) \sin 2A |r_p| |r_s| \sin(\delta - \Phi_{\text{sys}}),$$
(17)

$$I(2\Omega) = -I_0 \eta_2 I_2(\varphi_A) \sin 2(M - P) \begin{bmatrix} \sin 2M(|r_p|^2 \cos^2 A - |r_s|^2 \sin^2 A) \\ -\cos 2M|r_p||r_s| \sin 2A \cos(\delta - \Phi_{\rm sys}) \end{bmatrix}.$$
 (18)

*P*, *M*, and *A* are nominally set at 45°, 0°, and 135°, respectively, so that both the first harmonic and the second harmonic amplitude are non-zero. By adjusting *A* to null the first harmonic sets,  $A = 0^\circ$ . With  $A = 0^\circ$ , the second harmonic can be simplified as

$$I(2\Omega) = -I_0 \eta_2 J_2(\varphi_A) \sin 2(M - P) \sin 2M |r_p|^2.$$
(19)

By nulling the second harmonic, we find the setting for  $M = 0^{\circ}$ . Finally, keeping  $M = 0^{\circ}$ , the analyzer A is rotated to 45° to make both the first and second harmonic finite. By rotating P to null both harmonics, we find the setting for  $P = 0^{\circ}$ .

With  $P = 45^{\circ}$ ,  $M = 0^{\circ}$ , and  $A = 0^{\circ}$ , we inserted a phase shifter between the PEM and the sample. The first harmonic and the second harmonic in Eqs. (7) and (8) are now functions of the azimuth of the phase shifter *PS* and  $\Phi_{ps}$ ;

$$I(\Omega) = I_0 \eta_1 J_1(\varphi_A) |r_p|^2 \sin 2PS \sin \Phi_{\rm ps},$$
 (20)

$$I(2\Omega) = \frac{1}{2} I_0 \eta_2 J_2(\varphi_A) |r_p|^2 \sin 4PS(\cos \Phi_{\rm ps} - 1).$$
 (21)

By changing *PS* to null both harmonics, we find the setting for  $PS = 0^{\circ}$ . With *PS* away from 0°, we find the setting for  $\Phi_{ps} = 0^{\circ}$ .

After the calibration of the azimuths, one can then set  $P = 45^\circ$ ,  $PS = 0^\circ$ ,  $M = 0^\circ$ , and  $A = 135^\circ$  and have the first harmonic proportional to  $\sin(\delta - \delta_0)$  and the second harmonic proportional to  $\cos(\delta - \delta_0)$ . Deviations of *PS* and *M* from  $0^\circ$  introduce extra terms in both harmonics, making the absolute extraction of  $\sin(\delta - \Phi_{ps} - \Phi_{sys})$  and  $\cos(\delta - \Phi_{ps} - \Phi_{sys})$  from Eqs. (2) and (3) inaccurate. But if one only follows the changes in both harmonics as a result of surface processes, these extra terms do not have the effects if the deviations are small. We note that deviations of *P* and *A* from the specified values do

not have such effects. The azimuths of  $P = 45^{\circ}$  and  $A = 135^{\circ}$  only maximize the overall signals.

## 5. MEASUREMENT OF THE AMPLICATION FACTOR RATIO $\eta_1/\eta_2$

In an OI-RD system, the first harmonic (e.g., 50 kHz) and the second harmonics (e.g., 100 kHz) are measured with an electronic transducer that converts the photocurrent signal from a photodetector to a voltage signal, filters out the dc component, and amplifies the ac components for subsequent phase-sensitive detection with lock-in amplifiers. As shown in Eqs. (2) and (3), the amplification factor that includes the transmission of the cable between the transducer and the lock-in amplifier is frequency dependent,  $\eta_1$  for the first harmonic at 50 kHz and  $\eta_2$ for the second harmonic at 100 kHz. As a result,  $\eta_1/\eta_2$  needs to be measured for the accurate calculation of  $\delta - \delta_0$ . We measured  $\eta_1/\eta_2$  from the ratio of the first harmonic maximum  $(I_0\eta_1J_1(\varphi_A)|r_p||r_s|)$  to the second harmonic maximum  $(I_0\eta_2 J_2(\varphi_A)|r_p||r_s|)$  using a special value of modulation  $\varphi_A^J =$ 2.631 such that  $J_1(2.631) = J_2(2.631)$ . Then, we inserted a phase shifter (a Berek compensator, 5540, Newport, USA) between the PEM and the sample that adds an extra phase difference  $\Phi_{\rm DS}$ . As shown in Fig. 6, the first harmonic (filled triangles) and the second harmonic (empty circles) reach the maxima at respective  $\Phi_{ps}$ . From these maxima, we find  $\eta_1/\eta_2$  to be 1.36.

We define a total intensity maximum  $I_{max}$  as

$$I_{\max} = \sqrt{[I(\Omega)]^2 + (\eta_1/\eta_2)^2 [I(2\Omega)]^2} = I_0 \eta_1 J_1(\varphi_A) |r_p| |r_s|.$$
(22)

From Eq. (22),  $I_{\text{max}}$  should not be dependent on  $\Phi_{\text{ps}}$ . To check the validity of this expected result, we calculated  $I_{\text{max}}$  from the measured  $I(\Omega)$ ,  $I(2\Omega)$ , and  $\eta_1/\eta_2$  (from the ratio of the maxima of both harmonics) and displayed the result in Fig. 6 (open triangles) as well. The calculated  $I_{\text{max}}$  is indeed constant of  $\Phi_{\text{ps}}$  within 1%.



**Fig. 6.** Variation of the first harmonic amplitude  $I(\Omega)$ , second harmonic amplitude  $I(2\Omega)$ , and total intensity maximum  $I_{\text{max}}$  with phase difference  $\Phi_{\text{ps}}$  of the phase shifter.

#### 6. COMPARISONS OF FIRST HARMONIC NORMALIZATION METHODS

To deduce  $\Delta \delta = \delta - \delta_0$  from an OI-RD ellipsometry measurement using Eq. (4), there are two methods to determine the pre-factor  $I_0\eta_1 I_1(\varphi_A)$ . The first method, as used by Zhu and coworkers so far, divides the first harmonic in Eq. (4) by the maximum of Eq. (2)  $I_0\eta_1 J_1(\varphi_A)|r_{p0}||r_{s0}|$  on the bare substrate by adjusting the phase shifter  $\Phi_{ps}$ . The normalized first harmonic is  $|r_{b}||r_{s}|\sin(\delta-\delta_{0})/(|r_{b0}||r_{s0}|)$ . In the second method, as also used by Zhu and coworkers, the first harmonic in Eq. (4) is divided by the simultaneously measured second harmonic in Eq. (5) to yield  $tan(\delta - \delta_0)$ . When the molecular layer thickness *d* is much less than wavelength  $\lambda$ , both methods give the same results, since  $|r_p||r_s|$  is almost the same as  $|r_{p0}||r_{s0}|$  from the bare substrate. When  $|r_p||r_s|$  is not equal to  $|r_{p0}||r_{s0}|$  anymore, the second method provides a more accurate measurement of  $\Delta \delta = \delta - \delta_0$ , as demonstrated in the next paragraphs.

To compare these two normalization methods experimentally, we fabricated a microarray of biotin-labeled bovine serum albumin (BBSA) printed on epoxy-functionalized glass slides. At printing concentrations of 0.94, 1.88, 3.75, 7.5, 15, and 30  $\mu$ M, we printed four spots to form a row. The images of the microarray in the first harmonic and second harmonic of the OI-RD signal are shown in Fig. 7. For the second harmonic image, we see no noticeable features above the background when the printing concentration of the BBSA is equal to or less than 7.5  $\mu$ M, indicating that  $|r_{b}||r_{s}|$  from the BBSA-covered surface is almost identical to  $|r_{p0}||r_{s0}|$  from the bare surface. For printing concentrations larger than 7.5 µM, we clearly see darker spots of printed BBSA, which indicates that  $|r_p||r_s|$  from the BBSA-covered surface is less than  $|r_{p0}||r_{s0}|$  from the bare surface.

The first harmonic normalized against the maximum from the bare surface,  $|r_p||r_s|\sin(\delta-\delta_0)/(|r_{p0}||r_{s0}|)$ , and the first harmonic normalized against the simultaneously measured second harmonic,  $tan(\delta - \delta_0)$ , versus the BBSA printing concentration are plotted in Fig. 8.  $|r_p||r_s|\sin(\delta - \delta_0)/(|r_{p0}||r_{s0}|)$  is almost identical to  $tan(\delta - \delta_0)$  when the printing concentration of BBSA is equal to or less than 7.5  $\mu$ M. When the printing concentration further increases,  $tan(\delta - \delta_0)$  becomes larger

30µM

15µM

7.5µM

3.75µM

1.88µM

0.94µM





Fig. 8. Comparisons of the first harmonic amplitude normalized against the first harmonic maximum from a bare substrate  $|r_{p}||r_{s}|\sin(\delta-\delta_{0})/(|r_{b0}||r_{s0}|)$  with the first harmonic amplitude normalized against the simultaneously measured second harmonic amplitude  $tan(\delta - \delta_0)$ .

than  $|r_p||r_s|\sin(\delta-\delta_0)/(|r_{p0}||r_{s0}|)$ . This is expected as  $|r_p||r_s|$  is now smaller than  $|r_{p0}||r_{s0}|$  and shows that the second normalization method, having the first harmonic signal normalized against the simultaneously measured second harmonic, provides a more accurate detection of  $tan(\delta - \delta_0)$ .

#### 7. DISCUSSION AND CONCLUSIONS

As a form of ellipsometry, the OI-RD technique characterizes properties of a biomolecular layer on a transparent solid substrate by measuring  $\delta - \delta_0$ , which is related to the biomolecular layer thickness d or its surface mass density through Eq. (6). Such an optical characterization does not require biomolecules to be labeled with tags of any kind. The OI-RD ellipsometry has been used to both qualitatively and quantitatively characterize ultrathin films of solid materials and biomolecular layers (e.g., small organic molecules, DNA fragments, proteins and protein fragments, viruses, and bacteria) on solid surfaces with remarkable versatility and accuracy [13-20]. With polarizer azimuth P at 45°, PEM azimuth M at 0°, phase shifter azimuth PSat 0°, analyzer azimuth A at 135°, the OI-RD signals are related to  $\delta - \delta_0$  through Eqs. (3) and (4). To precisely determine  $\delta - \delta_0$ , the parameters of the polarizing optics need to be calibrated and set accurately.

The modulation amplitude  $\varphi_A$  of the PEM affects the amplitude of the first and the second harmonic signals, and in turn affects the accuracy of  $\delta - \delta_0$ . Currently, single-point calibration methods are widely used to calibrate the modulation amplitude  $\varphi_A$ . Some single-point methods are based on the observation of distinctive waveforms on the oscilloscope, some methods are based on finding set points where the Bessel functions produced by the harmonic signals are equal to zero. One limitation of the single-point methods is that only one calibration point may be obtained when distinctive waveforms appear or the Bessel functions cross zero. Therefore, a more encompassing calibration must be used to combine several single-point calibration methods together, where different measurement setups are usually involved. In this manuscript, we described a curve-fitting procedure to determine the modulation amplitude  $\varphi_A$  of a PEM in an OI-RD system, which could produce calibration curves over the entire usable range of the modulator. Compared to the single-point calibration method, the curve-fitting method is able to determine  $\varphi_A$  with an accuracy of 0.003 radians. We also showed that the static phase retardation  $\varphi_0$  can be determined by the ratio of a suitable first harmonic to the corresponding second harmonic of the OI-RD signal, and more importantly  $\varphi_0$  has no effect on the extraction of  $\delta - \delta_0$ .

The azimuths of polarization optics can cause significant deviation of OI-RD signals from  $\delta - \delta_0$ . With arbitrary azimuths for polarization optics, OI-RD signals are expressed as Eqs. (7) and (8), instead of Eqs. (3) and (4). We described a set of sequential procedures to precisely set azimuths of polarizer P, analyzer A, PEM M, and phase shifter PS. Deviations of PS and M away from ideally set values would add extra terms, making the absolute extraction of  $\delta$  –  $\Phi_{\rm ps}$  –  $\Phi_{\rm sys}$  inaccurate. However, if one only follows the change in the OI-RD signal after nulling, these extra terms do not affect the extraction of  $\delta - \delta_0$  when the deviations are small. The errors in the azimuths of P and A only affect the absolute amplitudes of the first and second harmonic signals so that the calculation of  $\delta - \delta_0$  is accurate after normalization. We explicitly show that by normalizing the first harmonic by the simultaneously acquired second harmonic gives a more accurate measurement of  $\delta - \delta_0$  than the method of normalizing the first harmonic signal by the maximum of first harmonic signal from the bare surface obtained by adjusting the phase shifter.

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